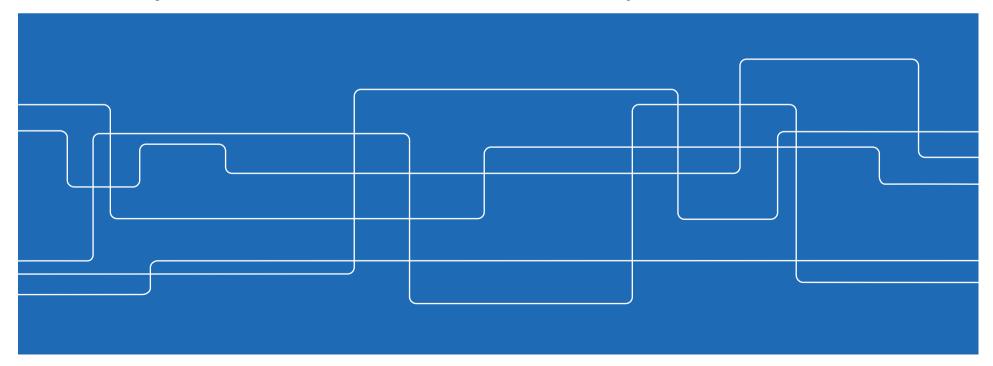


# **URLLC: System Design Perspectives through Queuing Analysis**

TUM URLLC Workshop, Zugspitze, July 2018 joint work with S. Schiessl, H. Al-Zubaidy





### **James Gross**



- Associate professor at KTH Stockholm (since 2012)
- Assistant professor at RWTH Aachen University (2008-12)
- PhD from TU Berlin in 2006
- Co-Founder of R3 Communications GmbH/Berlin
- Research focus:
  - Cellular networks
  - Critical machine-type communications
  - Theoretical network performance models
  - Edge computing and artificial intelligence

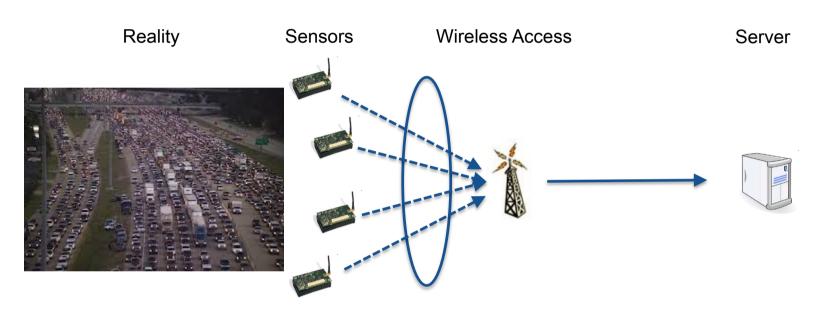


### **Outline**

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# **Machine-Type Communications: Origins**

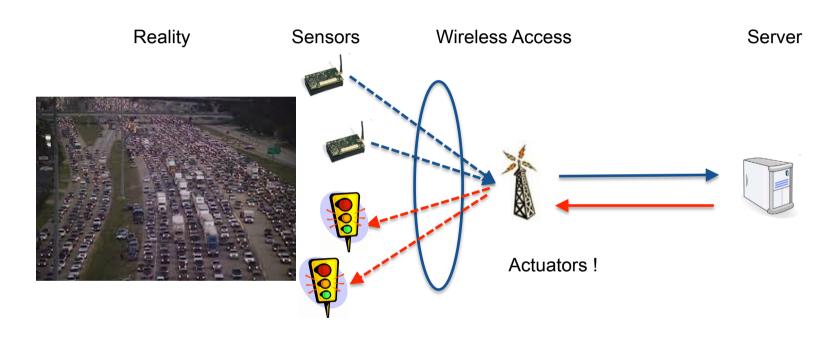


### <u>Autonomous</u> monitoring & metering purpose

- End of 90s: First research on "sensor networks"
- Mid 2000: First standards (802.15.4, 6LowPAN)
- ~2010: Picked up by cellular networking industry (M2M business)
  - → Massive machine-type communications



# Closing the Loop ...



- Closed-loop control (driven by autonomy trend)
- Dependability becomes the focus
  - → Critical machine-type communications!



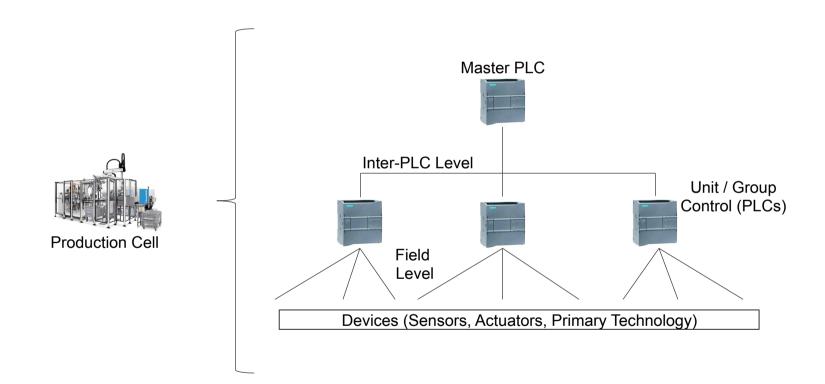
# **Critical MTC: Application Fields**

- Various application fields according to 3GPP [1]:
  - Rail-bound mass transit
  - Building automation
  - Factory of the future / industrial automation
  - Smart living / smarty city
  - Electric power distribution & power generation
- In addition:
  - Support for autonomous devices (cars, drones, robots)
  - Human-in-the-loop applications (AR / cognitive assistance)

3GPP, TR22.804 v1.0.0, December 2017



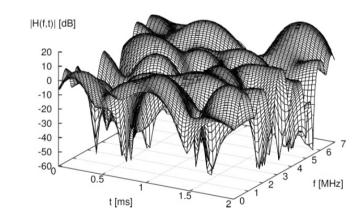
# **Critical MTC: Factory Automation**





# Range of Factory Automation Requirements

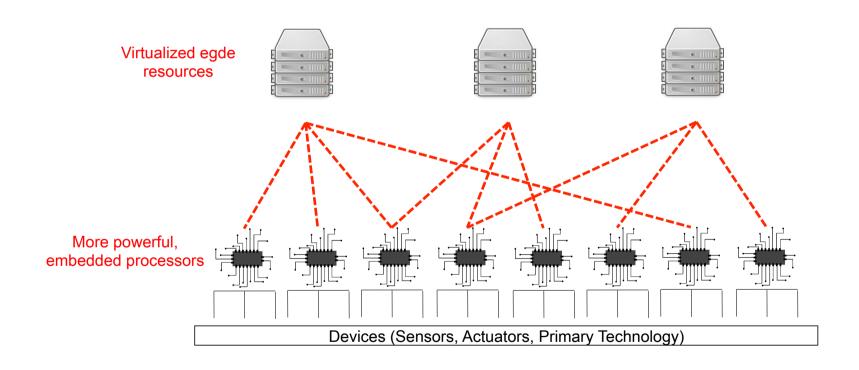
- Dependability: Availability + Reliability + Security
- Field-Level Control
  - Cycle time: <10 ms
  - Packet sizes: < 10 byte
  - Reliability: > 1 10<sup>-6</sup>
- Inter-PLC Communication:
  - Cycle time: < 50 ms
  - Packet sizes: < 500 byte</li>
  - Reliability: > 1 10<sup>-6</sup>



Why turn to wireless?

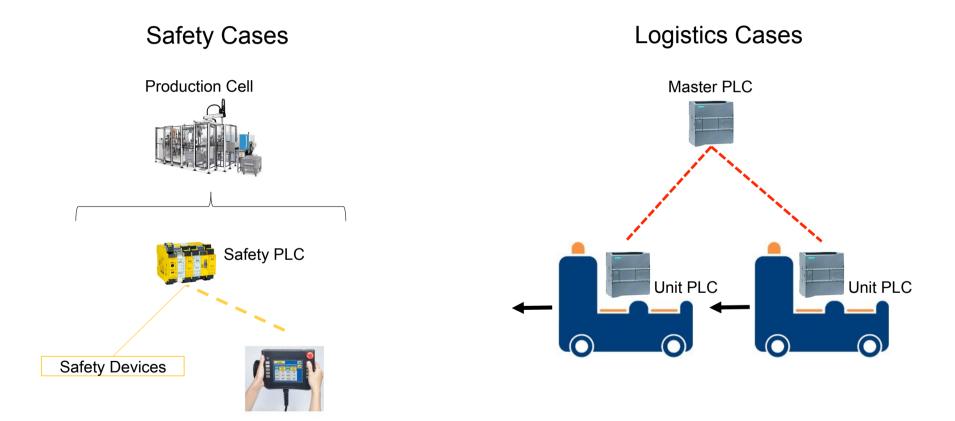


# **Visionary Reasoning: Flexibility**



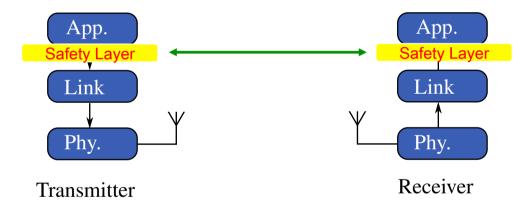


# Realistic Use Cases: Mobility-Driven





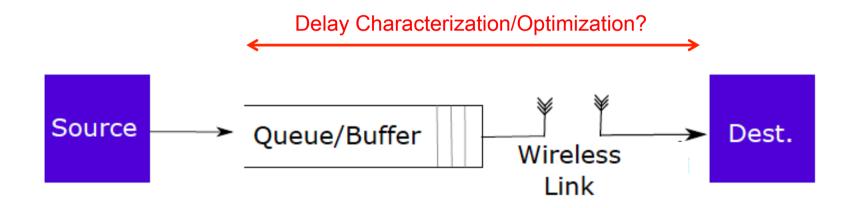
# **Systems & Safety Layers**



- Black channel principle
- Periodic message exchange, >10 ms cycle time
- Small PDUs, about 10 byte
- Turns link reliability issues into availability issues of the system



# **Queuing-Theoretic Problem Formulation**



- Deterministic arrivals
- Random service: Fading, interference, cross-traffic



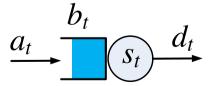
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# **Modeling Assumptions**

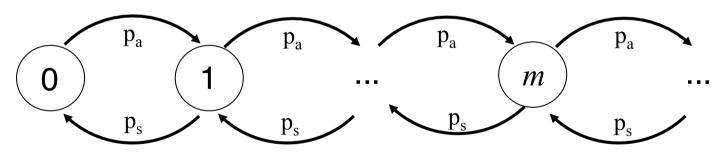
- Discrete time t
- Queue has infinite size
- Work-conserving server
- FIFO service order
- $a_t$ ,  $s_t$ ,  $d_t$ : Arrival, service and departure of slot t
- Arrival & service process are independent and stationary
- b<sub>t</sub>: Backlog at slot t





# **Traditional Approach: DTMCs**

- Per slot system size grows/decreases by 1, or stays the same
- Markov property of arrival and service process: With probability  $p_s$  system size decreases by 1 regardless of previous evolution ( $p_a$ : increases by 1)
  - → Homogeneous discrete-time birth-death Markov chain, steady state exists under certain conditions (stability criteria)



Steady-state analysis: 
$$\vec{\pi} = \vec{\pi} \cdot \mathbf{P}$$
 &  $\sum_{orall i} \pi^i = 1$ 



# **Traditional Approach: Pros & Cons**

Difference equation approach (balance equations)

#### Pros:

- 100 years of research: Lots of results, well understood
- Typically provides exact results

#### Cons:

- Simplicity hinges on Markov property / single packet event
- Quickly becomes intractable (concatenated systems, crosstraffic, scheduling)



### **Cumulative System View**

Define the following cumulative processes:

$$A_{s,t} = \sum_{i=s}^{t} a_i, \quad S_{s,t} = \sum_{i=s}^{t} s_i, \quad D_{s,t} = \sum_{i=s}^{t} d_i$$

Let us assume that new arrivals can be served instantly.

Denote the backlog at time t as  $b_t$ , we have (Lindley):

$$b_t = \max(0, b_{t-1} + a_t - s_t)$$

As the system is lossless, we also have:

$$b_t = A_{0,t} - D_{0,t}$$



# **Exercise: From Lindley to Reich!**

Work through the recursion of Lindley's equation (use  $b_0 = 0$ )

$$b_{t} = \max(0, b_{t-1} + a_{t} - s_{t})$$

$$= \max(0, \max(0, b_{t-2} + a_{t-1} - s_{t-1}) + a_{t} - s_{t})$$

$$= \max(0, \max(a_{t} - s_{t}, b_{t-2} + a_{t} + a_{t-1} - s_{t} - s_{t-1}))$$

$$= \max(0, A_{t,t} - S_{t,t}, b_{t-2} + A_{t-1,t} - S_{t-1,t})$$

$$= \max_{0 \le i \le t} (0, A_{i,t} - S_{i,t})$$

$$= \max_{0 \le i \le t} (A_{i,t} - S_{i,t})^{+}$$



# Min,+ System Theory of Queuing Systems

What does Reich's equation mean for the system output?

$$b_{t} = A_{0,t} - D_{0,t} \Leftrightarrow D_{0,t} = A_{0,t} - b_{t}$$

$$= A_{0,t} - \max_{0 \le i \le t} (A_{i,t} - S_{i,t})^{+}$$

$$= \min_{0 \le i \le t} (A_{0,t} - A_{i,t} + S_{i,t})$$

$$= \min_{0 \le i \le t} (A_{0,i-1} + S_{i,t})$$

$$= (A \oplus S)_{0,t}$$

Turns out that:  $b_t = A_{0,t} - D_{0,t}$   $= \max_{0 \le i \le t} (A_{i,t} - S_{i,t})^+$  $= (A \ominus S)_{t,t}$ 

with: 
$$(X \ominus Y)_{s,t} = \max_{\tau \le s} (X_{\tau,t} - Y_{\tau,s})$$



# **Probabilistic Backlog Bound**

#### First consider:

$$\mathbb{P}\left((X\ominus Y)_{s,t} \geq z\right) = \mathbb{P}\left(\max_{\tau \leq s} \left(X_{\tau,t} - Y_{\tau,s}\right) \geq z\right)$$
Union Bound
$$\leq \sum_{\tau=0}^{s} \mathbb{P}\left(X_{\tau,t} - Y_{\tau,s} \geq z\right)$$
Chernoff
Bound
$$\leq e^{-\theta z} \cdot \sum_{\tau=0}^{s} \mathbb{M}_{X}(\theta, \tau, t) \cdot \mathbb{M}_{Y}(-\theta, \tau, s)$$

$$= \epsilon$$

#### Thus:

$$\mathbb{P}\left((A \ominus S)_{t,t} \ge \max_{0 \le \theta} \left(\frac{1}{\theta} \left(\log \sum_{\tau=0}^{t} \mathbb{M}_{A}(\theta, \tau, t) \cdot \mathbb{M}_{S}(-\theta, \tau, t) - \log \epsilon\right)\right)\right) \le \epsilon$$



### **Stochastic Network Calculus: Pros & Cons**

Moment-bounds on system variables

#### Pros:

- Applicable for arbitrary arrival and service processes
- Strict upper bound on system performance
- Works also for concatenated systems

#### Cons:

- Best for stationary processes with independent increments
- Upper bound is not tight in general

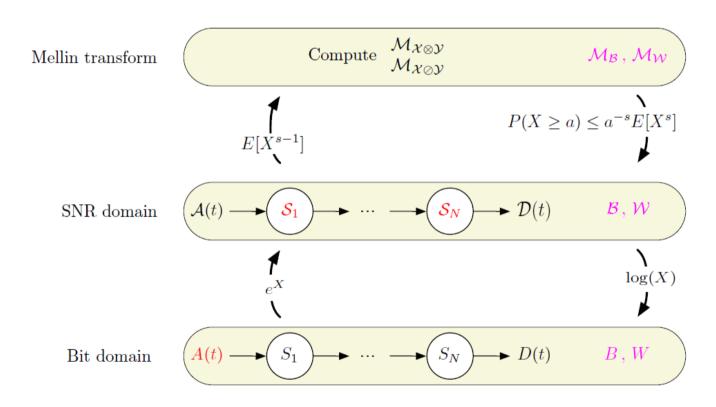


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### From Bit-Domain SNC to SNR-Domain SNC



H. Al-Zubaidy et al. "Network-layer Performance Analysis of Multi-hop Fading Channels," Transactions on Networking, 24/1, 2016



### SISO Interference Channel

Signal-of-interest and interference signals are fading.

$$\gamma_t = \frac{P_0 |h_{0,t}|^2}{\sum_i P_i |h_{i,t}|^2 + \sigma^2}$$

Service in time slot *t* in bits:

$$S_t = n \log_2(1 + \gamma_t)$$

w.l.o.g., assume n/log(2)=1.

→ Service in the **SNR-domain**:

$$S_t = e^{S_t} = 1 + \gamma_t$$



### SISO Interference Channel

For the queueing analysis, we must find

$$\mathcal{M}_{\mathcal{S}}(\theta) = \mathbb{E}\left[\mathcal{S}^{\theta-1}\right] = \int_{0}^{\infty} (1+\gamma)^{\theta-1} f(\gamma) d\gamma$$

For K interferers, we get K integrals of the form

$$\int_0^\infty \frac{(1+\gamma)^{\theta-2}}{\gamma+a} e^{-\gamma} d\gamma = \int_1^\infty \frac{z^{\theta-2}}{z+a-1} e^{-z+1} dz$$

S. Schiessl et al. "On the Delay Performance of Interference Channels," IFIP Networking, 2016.



### **SISO Interference Channel**

#### Solution:

- Split the integral into two parts: z < a-1 and z > a-1
- For the second part with z > a-1:

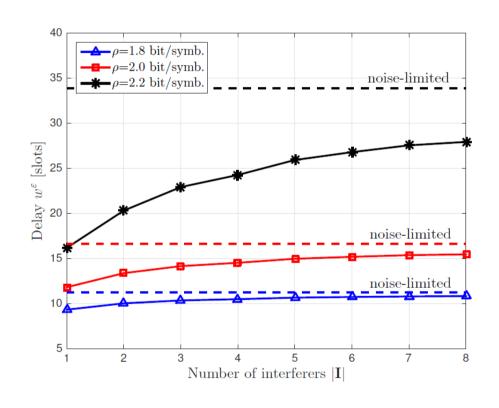
$$\frac{z^{\theta-3}}{1 + \frac{a-1}{z}} = z^{\theta-3} \sum_{n=0}^{\infty} \left(\frac{1-a}{z}\right)^n$$

- For the first part: similar solution
- $\rightarrow$  Can determine  $\mathcal{M}_{S}(\theta)$  in closed form (as a series of incomplete gamma functions)

S. Schiessl et al. "On the Delay Performance of Interference Channels," *IFIP Networking*, 2016. F. Naghibi et al. "Performance of Wiretap Rayleigh Fading Channels under Statistical Delay Constraints," *IEE ICC*, 2017



### SISO Interference Channel: Main Result



- Main av. signal power: 15 dB
- Total av. interference power is constant (8 dB).
- What is the max. delay w such that  $p_{v}(w) < 10^{-6}$ ?

#### Result:

It is better to have one interf. with av. P=8 dB than two interf. with av. P=5 dB each.

Reason: signal from the one interferer is often weak, allowing high data rates

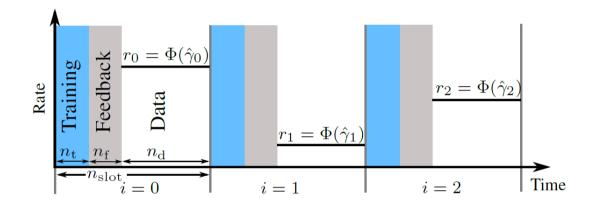


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### Finite Blocklength and Imperfect CSIT



- SISO set-up, focus on impact of CSI at transmitter:
  - Trade-off 1: Training symbols  $n_{\rm t} \Leftrightarrow$  Data symbols  $n_{\rm d}$
  - Trade-off 2: Rate r ⇔ Error probability ε
  - $\rightarrow$  Errors are bad, but low r and small  $n_{\rm d}$  can also increase the queueing delay!



# Finite Blocklength and Imperfect CSIT

Normal approximation (Polyanskiy et al. / Yang et al.):

$$\varepsilon \approx \mathbb{E}\left[\left.Q\left(\frac{\log_2(1+\Gamma)-r}{\sqrt{\mathcal{V}(\Gamma)/n_{\rm d}}}\right)\right| \hat{\gamma}\right] \qquad \begin{array}{c} \Gamma : \text{Actual SNR} \\ \text{(unknown/random)} \\ \hat{\gamma} : \text{Estimated SNR} \end{array}\right.$$

### Too complex for queueing analysis.

Thus, we find a normal approximation for  $\Gamma$  and use a Taylor approximation for the FBL effects, giving:

$$\varepsilon \approx Q \left( \frac{\hat{\gamma} - (2^r - 1)}{\sigma_{\text{ICSI,FBL}}} \right)$$

S. Schiessl et al. "Delay Performance of Wireless Communications with Imperfect CSI and Finite Length Coding," accepted for publication Transactions on Communications, 2018.



# Finite Blocklength and Imperfect CSIT

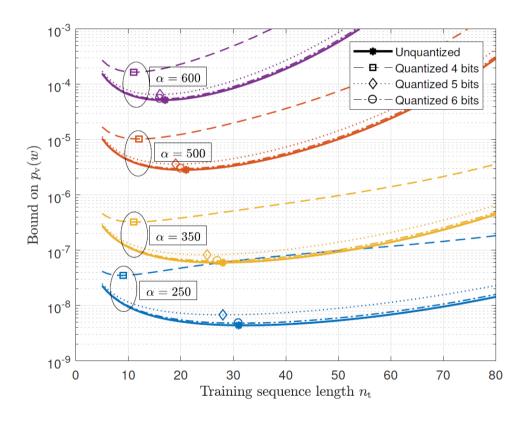
To minimize the delay violation probability, minimize

$$\mathcal{M}_{\mathcal{S}}(\theta) = \mathbb{E}\left[\mathcal{S}^{\theta-1}\right] = \int_{0}^{\infty} (1+\gamma)^{\theta-1} f(\gamma) d\gamma$$

- For each estimated SNR  $\hat{\gamma}$ : need to solve trade-off  $r \Leftrightarrow \varepsilon$
- Can be solved quickly, as the expression is convex in the approximate  $\varepsilon$



# Main Result 1: Optimal $n_t$

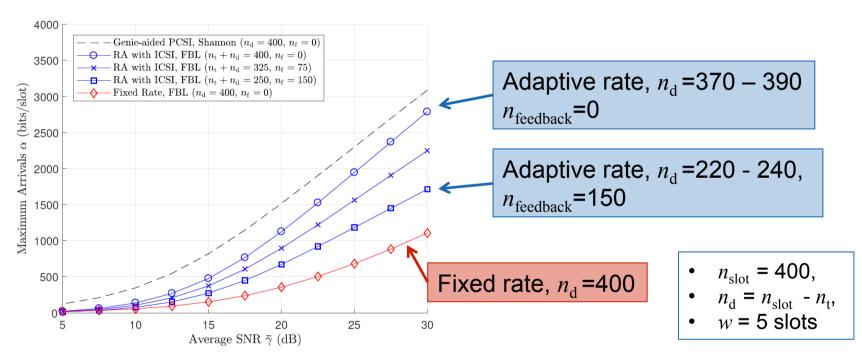


### Parameters:

- $n_{\rm slot} = 250$ ,
- $n_{\rm d} = n_{\rm slot} n_{\rm t},$  w = 5 slots,
- Avg. SNR 15 dB



# **Result 2: Rate Adaptation is Superior**



- These results consider queueing constraints:  $p_v(w=5) < 10^{-8}$
- Ignoring the queueing constraints would lead to wrong conclusions.



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### **Multiuser MISO**

Multiuser MISO with zero-forcing beamforming (ZFBF). M antennas, K scheduled users

Large K: multiplexing gain



Small K: beamforming gain

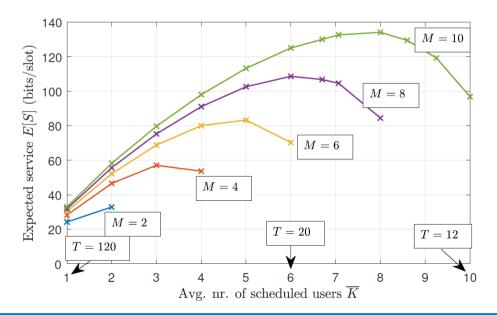
What is the optimal K under delay constraints?

S. Schiessl et al. "On the Delay Performance of the Multi-user MISO Downlink," ArXiv preprint, 2018.



### **Multiuser MISO**

- Has been well studied with respect to ergodic sum rate, e.g., Hochwald & Vishwanath '02.
- Choose K≈αM. Here: α ≈ 0.8

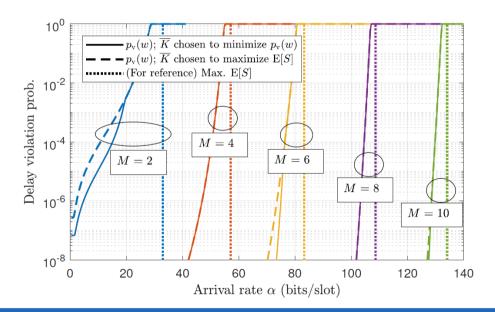


- $n_{\text{slot}} = 400$ ,
- $K_{\text{tot}}$  = 120 users,
- $P_{\text{sum}}^{\text{sum}}$  = 20 dB



# **Multiuser MISO: Delay Performance**

- Observation: For M ≥ 6, no queueing delay as long as expected arrival rate < 0.9 \* expected service rate</li>
- Optimal value K rarely changes under delay constraints



- $n_{\rm slot} = 400$ ,
- $K_{\text{tot}}$  = 120 users,
- $P_{\text{sum}} = 20 \text{ dB},$
- w = 120 slots

FYI: When K=2, each of the  $K_{\text{tot}}$ =120 users can be scheduled 2 times within w=120 slots.



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### **Discussion**

- Queuing analysis extends physical layer work towards real application layer performance
- SNC approaches can provide useful upper bounds
- Somewhat surprising findings for URLLC:
  - Have rather one strong interferer
  - Estimate channel & rate adaption
  - Relatively few antennas at transmitter lead (through channel hardening) already to almost perfect system performance



### **Outlook**

- Transient system characterization instead of steady-state
- Analyze the entire loop through edge server
- Integrate models with control performance models